

Today, we will:

- Do some example problems – Filters
- Review the pdf module: **Digital Filters**

Example: Low-pass filter circuit

Given: The following components are available:

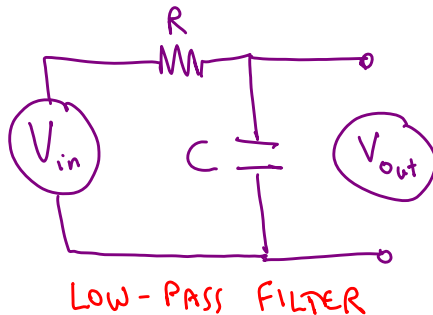
- two resistors, 10.0 k Ω and 87.0 k Ω
- a capacitance decade box



The input voltage contains a signal frequency of 50 Hz, and some unwanted noise at 1000 Hz. We want to amplify the signal, with $G_{\text{amplifier}} \approx 8.5$, and we decide to use a first-order low-pass filter with a cutoff frequency of 200 Hz to attenuate the noise.

(a) To do: Draw the filter circuit and calculate the required capacitance.

Solution:



$$f_{\text{cutoff}} = \frac{1}{2\pi RC} \rightarrow \text{solve for } C:$$

$$C = \frac{1}{2\pi R f_{\text{cutoff}}}$$

Let's pick the 87.0 k Ω resistor

$$C = \frac{1}{2\pi (87,000 \cancel{\Omega}) (200 \cancel{\text{Hz}})} \left(\frac{\cancel{\text{Coulomb}}}{\cancel{\text{S} \cdot \text{A}}} \right) \left(\frac{\cancel{\text{A}}}{\cancel{\text{V}}} \right) \left(\frac{\cancel{\text{F} \cdot \text{V}}}{\cancel{\text{C}}} \right)$$

Unity conversion factors

$$\therefore \text{So, } C = 9.1468 \times 10^{-9} \text{ F} \rightarrow C = 0.00915 \mu\text{F}$$

[In real life, a capacitor of 0.01 μF is close enough & this is a common capacitance]

(b) To do: Calculate the overall gain of both the signal and the noise, and calculate how well we have improved the signal-to-noise ratio (SNR). Note: **SNR is the ratio of signal power to noise power**. Since electrical power is proportional to voltage², we define SNR as

$$\text{SNR} = \frac{\text{Power}_{\text{signal}}}{\text{Power}_{\text{noise}}} = \left(\frac{|V_{\text{signal}}|}{|V_{\text{noise}}|} \right)^2, \text{ or in terms of decibels, } \text{SNR}_{\text{dB}} = 20 \log_{10} \left(\frac{|V_{\text{signal}}|}{|V_{\text{noise}}|} \right).$$

Solution:

Signal: $f = 50 \text{ Hz} \rightarrow G_{\text{overall}} = G_{\text{amp}} \cdot G_{\text{filter}}$



recall, for a 1st-order low-pass filter,

$$G = \frac{1}{\sqrt{1 + \left(\frac{f}{f_{\text{cutoff}}} \right)^2}}$$

$$= (8.50) \frac{1}{\sqrt{1 + \left(\frac{50}{200} \right)^2}}$$

(given)

$$G_{\text{overall}} = 8.2462$$

This is the overall gain of the signal (the 50 Hz component)

• Repeat for the noise @ $f = 1000 \text{ Hz}$

$$G_{\text{overall}} = G_{\text{amp}} G_{\text{filter}} = (8.50) \frac{1}{1 + \left(\frac{1000}{200}\right)^2} = \underline{\underline{1.6670}}$$

★ Notice → G_{overall} of signal
is $>$ G_{overall} of noise,
which is what we desire → the filter will improve the signal-to-noise ratio

This is the overall gain
of the noise

• Compare the original SNR to the signal-conditioned (i.e., low-pass filtered) SNR.

$$(SNR)_{\text{original}} = \left(\frac{|V|_{\text{signal}}}{|V|_{\text{noise}}} \right)^2$$

$$(SNR)_{\text{filtered}} = \left(\frac{8.2462 |V|_{\text{signal}}}{1.6670 |V|_{\text{noise}}} \right)^2$$

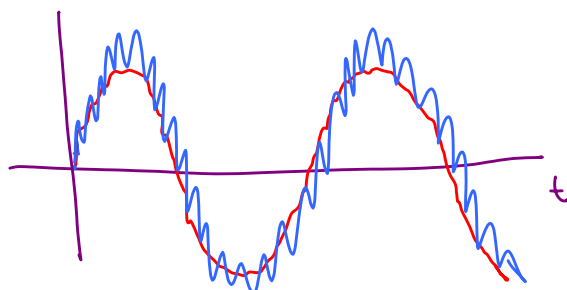
The ratio of these is the improvement.

Note: We are not given the values of $|V|_{\text{signal}}$ & $|V|_{\text{noise}}$, but they cancel out when we calculate the ratio.

$$\frac{(SNR)_{\text{filtered}}}{(SNR)_{\text{original}}} = \left(\frac{8.2462}{1.6670} \right)^2 = 24.47 \approx 24.5$$

★ So, we have improved the SNR by a factor of 24.5 by putting a low-pass filter in line between the amplifier & the DAQ

Bottom line: When there is unwanted high frequency noise, we can improve the signal-to-noise ratio by using a low-pass filter, and the filter also reduces the potential of aliasing



— Unfiltered (has high-frequency noise)

— filtered (we have reduced or attenuated the high-frequency noise
[but not completely])

Example: Digital data acquisition and Filters

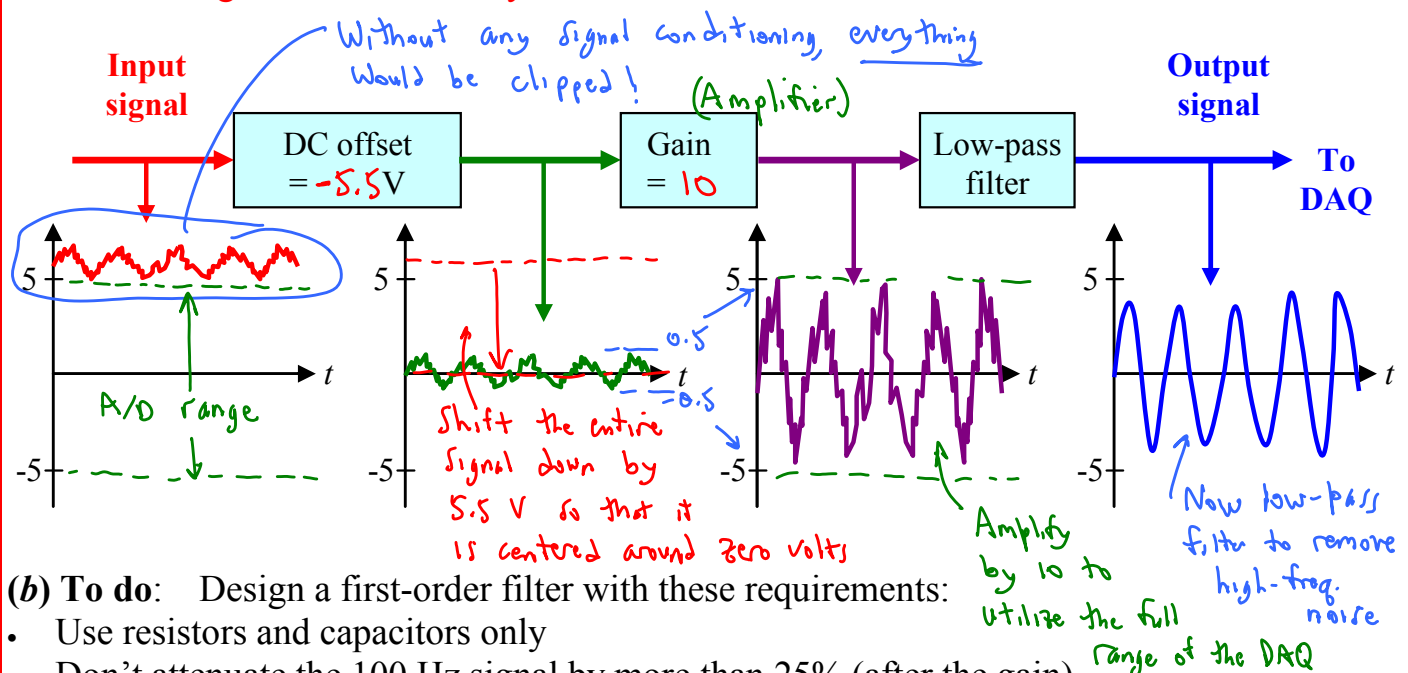
Given: A voltage signal has a frequency around 100 Hz, and ranges from 5.0 to 6.0 volts. There is also some unwanted AC noise at a frequency around 5000 Hz, with an amplitude of ± 0.005 V. The digital data acquisition system is 12-bit, and has a range from -5 to 5 volts.

(a) To do: Calculate an appropriate DC offset and gain in order to utilize the full range of the A/D system.

Solution: $\text{DC offset} = -5.5 \text{ V}$, $\text{Gain} = 10$

[We will learn later how to build these circuits (DC offset & amplifier)]

Schematic diagram of the circuitry:



(b) To do: Design a first-order filter with these requirements:

- Use resistors and capacitors only
- Don't attenuate the 100 Hz signal by more than 25% (after the gain).
- Attenuate the 5000 Hz noise to lower than the quantizing error of the A/D system (in other words, remove it completely, since its amplitude will be reduced to the noise level of the A/D converter).

Solution: (a) See answers and discussions above.

✱ There are two problems with our original setup:

(1) Signal is clipped (all our voltages are $> 5 \text{ V}$, so they would be clipped by the -5 to 5 V A/D converter)

(2) Even after we shift the signal down to avoid clipping, we are using a small portion of the A/D range

(-0.5 to 0.5 V = 1 V range out of -5 to 5 V = 10 V A/D)

[We are using only $1/10^{\text{th}}$ of the available range of the A/D!]

(b)

• Low-pass filter:

- First criterion — use R & C only (passive 1st-order low-pass filter)
- Now add a low-pass filter to remove the high frequency noise.
- Calculate the minimum cutoff frequency such that the second criterion is met → the signal @ 100 Hz is attenuated by not more than 25% → Set $G \geq 0.75$ @ 100 Hz.

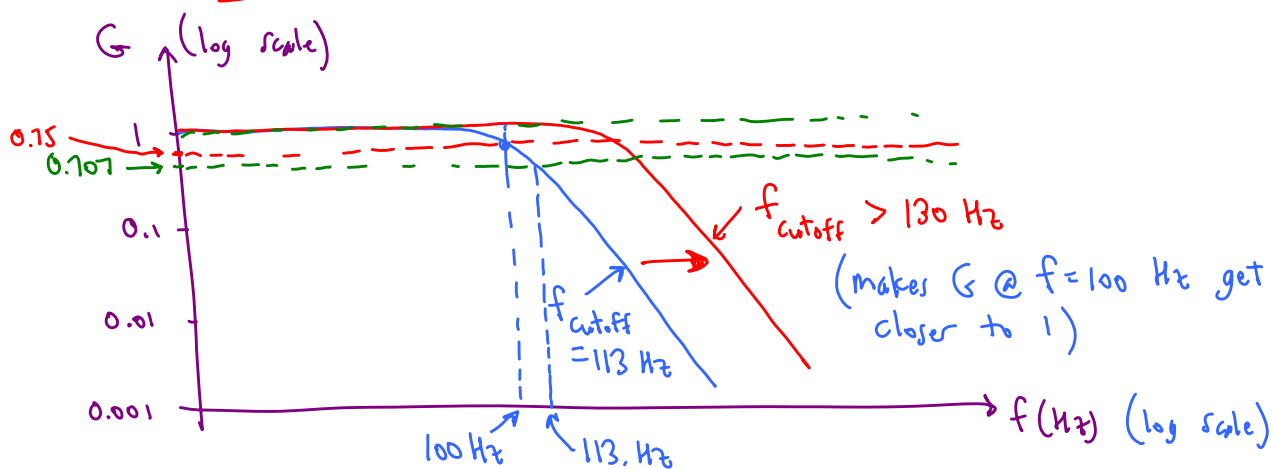
• For a first-order filter, $n=2$,
$$G = \frac{1}{\sqrt{1 + \left(\frac{f}{f_{\text{cutoff}}}\right)^2}}$$

- Solve for f_{cutoff} →
(leaving out the algebra)

$$f_{\text{cutoff}} = \frac{f}{\sqrt{\frac{1}{G^2} - 1}}$$

• Plug in $f = 100 \text{ Hz}$ & $G = 0.75 \rightarrow f_{\text{cutoff}} = \frac{100 \text{ Hz}}{\sqrt{\frac{1}{(0.75)^2} - 1}} = \underline{\underline{113. \text{ Hz}}}$

This value of $f_{\text{cutoff}} = 113. \text{ Hz}$ is the lower limit, because any $f_{\text{cutoff}} > 113 \text{ Hz}$ will shift the Bode plot to the right & will make $G > 0.75$, which would be good.



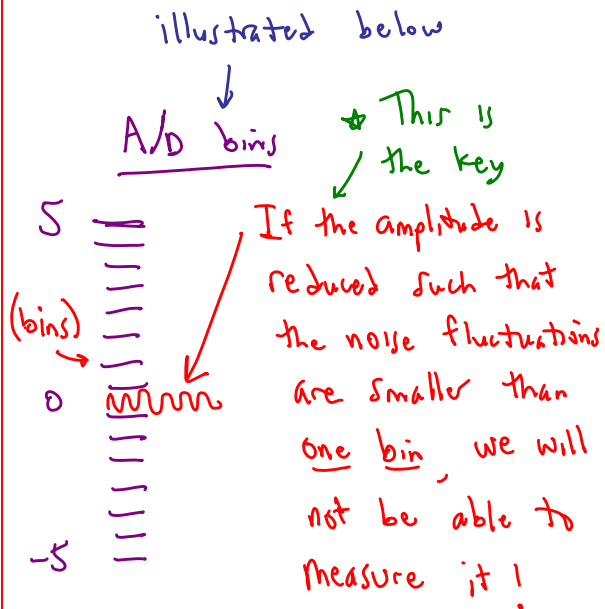
• Third Criterion $\rightarrow$ Remove the high frequency noise completely

• Recall, Quantizing error of the A/D converter $= \pm \frac{V_{\max} - V_{\min}}{2^{N+1}}$

Here, A/D is $N=12$ bit & range is $V_{\min} \rightarrow -5$ to $V_{\max} \rightarrow 5$ V

$$\text{So, } \underline{\text{quantizing error}} = \pm \frac{5 - (-5)V}{2^{12+1}} = \pm \underline{\underline{0.00122 \text{ V}}}$$

• Our goal is to attenuate the 5000 Hz noise amplitude so much that it gets "lost" in the quantizing error of the A/D, as illustrated below



• So, now we calculate the upper limit of f_{cutoff} that will kill the noise as shown to the left

• Caution $\rightarrow$ The noise amplitude is ± 0.005 V, but we amplify by 10.

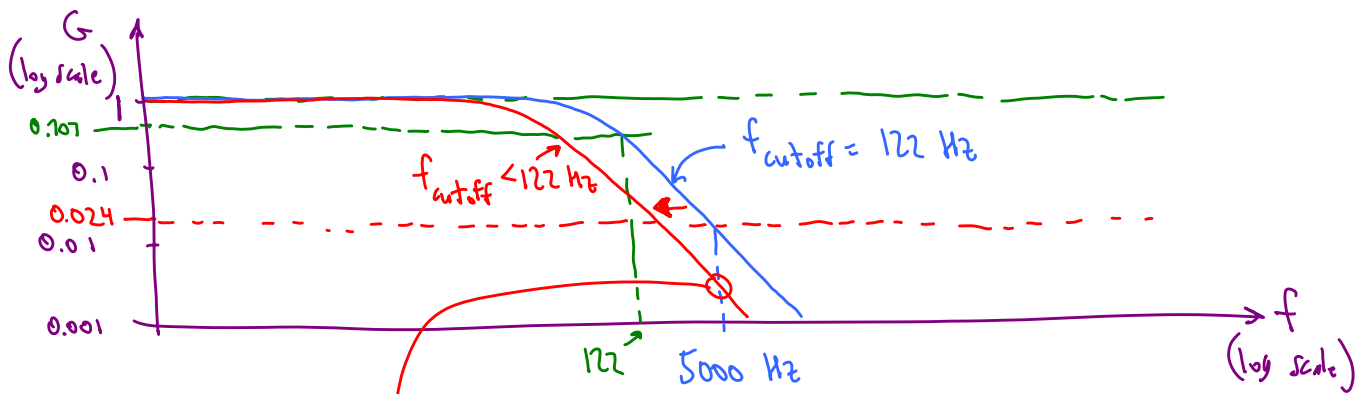
So, the amplitude of the noise going into the A/D is not ± 0.005 V, but is $\pm \underline{\underline{0.05 \text{ V}}}$

• Set $G = \left| \frac{V_{\text{out}}}{V_{\text{in}}} \right| = \left| \frac{\pm 0.00122 \text{ V}}{\pm 0.05 \text{ V}} \right| = \underline{\underline{0.0244}}$ = minimum G to kill all the noise

Quantizing error (pointing to 0.00122 V)
Amplitude of the noise (pointing to 0.05 V)

• Calculate $f_{\text{cutoff}} = \frac{f}{\sqrt{\frac{1}{G^2} - 1}} = \frac{5000 \text{ Hz}}{\sqrt{\frac{1}{0.0244^2} - 1}} = \underline{\underline{122 \text{ Hz}}}$

★ This value of f_{cutoff} is the upper limit, since any $f_{\text{cutoff}} < 122 \text{ Hz}$ would attenuate the noise even more!

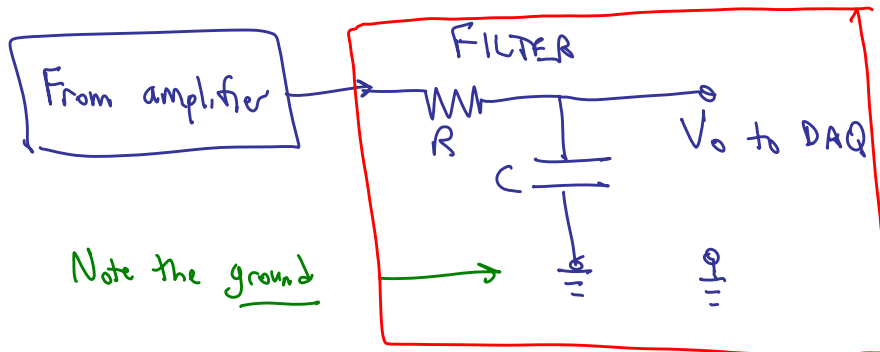


★ Any $f_{\text{cutoff}} < 122 \text{ Hz}$ shifts the Bode plot to the left ∴ G at $f = 5000 \text{ Hz}$ would be even smaller than the requirement

• Bottom line → need $113 < f_{\text{cutoff}} < 122 \text{ Hz}$ to meet all the criteria

• We pick f in the middle, i.e. pick $f_{\text{cutoff}} = 117 \text{ Hz}$

• Circuit: Low-pass filter with $f_{\text{cutoff}} = 117 \text{ Hz}$:



• Pick a capacitor → $C = 0.100 \mu\text{F}$ is a popular one. Thus,

$$R = \frac{1}{2\pi f_{\text{cutoff}} C} = \frac{1}{2\pi (117 \frac{1}{\text{s}}) (0.100 \times 10^{-6} \text{F})} \underbrace{\left(\frac{\text{F} \cdot \cancel{\text{V}}}{\cancel{\text{C}}} \right) \left(\frac{\cancel{\text{C}}}{\cancel{\text{s}} \cdot \text{A}} \right) \left(\frac{\cancel{\text{N}} \cdot \text{A}}{\cancel{\text{V}}} \right)}_{\text{Unity conversion factors}}$$

$$R = 13,603 \Omega, \text{ or } R = 13.6 \text{ k}\Omega$$

• Final answer $\left. \begin{array}{l} R = 13.6 \text{ k}\Omega \\ C = 0.100 \mu\text{F} \end{array} \right\}$ These values will do the job for us!